

# ST777/778/779

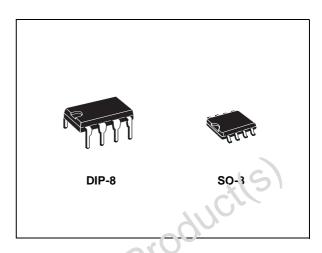
## LOW VOLTAGE INPUT, 3-3.3V/5V/ADJUSTABLE OUTPUT DC-DC CONVERTER WITH SYNCHRONOUS RECTIFIER

- 1V TO 6V INPUT GUARANTEES START-UP UNDER LOAD
- MAXIMUM OUTPUT CURRENT OF 300mA (778 OR 779 ADJUSTED TO 3V)
- LOAD FULLY DISCONNECTED IN SHUTDOWN
- TYPICAL EFFICIENCY OF 82%
- INTERNAL 1A POWER SWITCH AND SYNCHRONOUS RECTIFIER
- ADJUSTABLE CURRENT LIMIT ALLOWS LOW-COST INDUCTORS
- SUPPLY CURRENT OF 270µA (NO LOAD)
- SHUTDOWN SUPPLY CURRENT 20µA
- PACKAGE AVAILABLE: DIP-8 AND SO-8

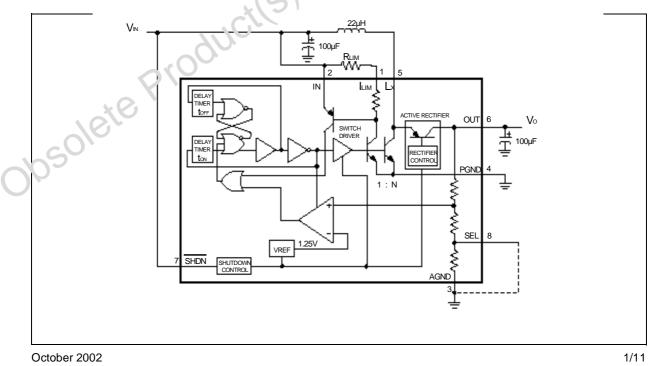
## DESCRIPTION

The ST777/778/779 are dc-dc converters that step-up from low voltage inputs requiring only three external components, an inductor (typically  $22\mu$ H) and two capacitors. The device include a Sinchronous Rectifier that eliminates the need for an external catch diode, and allows regulatior even when the input is greater than the output.

## SCHEMATIC DIAGRAM



Unlike others step up DC-DC converters the ST777/778/779's Sin chronous Rectifier turns off in the shutdow, node, fully disconnecting the output from the source. This eliminates the current drain associated with conventional step-up converters when off or in shutdown. Supply current is 270µA under no load and only 20µA in stand by mode.



### **ABSOLUTE MAXIMUM RATINGS**

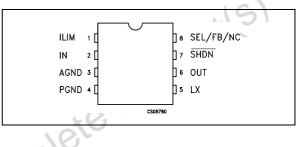
Symbol	Parameter	Value	Unit
V <sub>CC</sub>	DC Input Voltage to GND	-0.3 to +7	V
LX	Switch off Pin Voltage	-0.3 to +7	V
LX	Switch on Pin Voltage	30 sec short to IN or OUT	
OUT, SHDN	Output, Shutdown Voltage	-0.3 to +7	V
AGND to PGND	Analog and Power Ground	-0.3 to +0.3	V
FB	FB Pin Voltage	-0.3 to (OUT+0.3)	V
P <sub>TOT</sub>	Continuous Power Dissipation (at $T_A = 85^{\circ}C$ ) DIP-8	550	mW
	Continuous Power Dissipation (at $T_A = 85^{\circ}C$ ) SO-8	344	TIVV
T <sub>STG</sub>	Storage Temperature Range	-40 to 150	°C
T <sub>OP</sub>	Operating Ambient Temperature Range	0 to 85	°C

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these condition is not implied.

## **ORDERING CODES**

TYPE	DIP-8	SO-8
ST777	ST777ACN	ST777ACD
ST778	ST778ACN	ST778ACD
ST779	ST779ACN	ST779ACD

## **CONNECTION DIAGRAM**



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#### **PIN CONNECTIONS**

Pin No.	SYMBOL	NAME AND FUNCTION
1	ILIM	Sets switch current limit input. Connect to IN for 1A current limit. A resistor from ILIM to IN sets lower peak inductor currents.
2	IN	Input from battery
3	AGND	Analog ground. Not internally connected to PGND.
4	PGND	Power ground. Must be low impedance; solder directly to ground plane or star ground. Connect to AGND, close to the device.
5	LX	Collector of 1A NPN power switch and emitter of Sinchronous Rectifier PNP.
6	OUT	Voltage Output. Connect filter capacitor close to pin.
7	SHDN	Shutdown input disables power supply when low. Also disconnets load from input. Threshold is set at $V_{\text{IN}}/2$ .
8	SEL/N.C./FB	<ul> <li>Selection pin for 3/3.3V version (778);</li> <li>Not internally connected for 5V version (777);</li> <li>Feedback pin for adjustable version (779).</li> </ul>

## THERMAL DATA

Symbol	Parameter	DIP-8	SO-8	Unit
R <sub>thj-amb</sub>	Thermal Resistance Junction-ambient	100	160	°C/W

**ELECTRICAL CHARACTERISTICS** (V<sub>IN</sub>=2.5V, C<sub>I</sub> = 22µF, C<sub>O</sub>=100µF, SHDN and ILIM connected to IN, AGND connected to PGND, T<sub>A</sub>=0 to 85°C, unless otherwise specified. Typical values are referred at T<sub>A</sub>=25°C)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V <sub>START</sub>	Start up Voltage	$I_{LOAD}$ < 10mA, $T_A$ = 25°C (Note 1)			1	V
V <sub>IN(MAX)</sub>	Maximum Input Voltage	(Note 1,2)	6			V
V <sub>O</sub>	Output Voltage ST777 779 (set to 5V), (Note 3)	$\begin{split} & I_{LOAD} \leq 30 \text{mA}, \ V_{IN} = 1.1 \text{V to 5V or} \\ & I_{LOAD} \leq 80 \text{mA}, \ V_{IN} = 1.8 \text{V to 5V or} \\ & I_{LOAD} \leq 130 \text{mA}, \ V_{IN} = 2.4 \text{V to 5V} \end{split} $		5.0	5.2	V
	Output Voltage ST778 (Note 3)	$\begin{split} & \text{SEL=0V} \\ & \text{I}_{\text{LOAD}} \leq 50 \text{mA}, \text{V}_{\text{IN}} = 1.1 \text{V to } 3.3 \text{V or} \\ & \text{I}_{\text{LOAD}} \leq 210 \text{mA}, \text{V}_{\text{IN}} = 1.8 \text{V to } 3.3 \text{V or} \\ & \text{I}_{\text{LOAD}} \leq 300 \text{mA}, \text{V}_{\text{IN}} = 2.4 \text{V to } 3.3 \text{V} \end{split}$	3.17	3.30	3.43	V
		$\label{eq:selection} \begin{split} & \text{SEL=OPEN} \\ & \text{I}_{\text{LOAD}} \leq 30 \text{mA}, \ \text{V}_{\text{IN}} = 1.1 \text{V to } 3 \text{V or} \\ & \text{I}_{\text{LOAD}} \leq 210 \text{mA}, \ \text{V}_{\text{IN}} = 1.8 \text{V to } 3 \text{V or} \\ & \text{I}_{\text{LOAD}} \leq 300 \text{mA}, \ \text{V}_{\text{IN}} = 2.4 \text{V to } 3 \text{V} \end{split}$	2.88	3.00	3.12	V
	Output Voltage Range ST779	(Note 4)	2.7	. (	6.5	V
I <sub>IN</sub>	No Load Supply Current	I <sub>LOAD</sub> = 0 mA, (Switch ON) (Note 5)		270		μA
I <sub>SHDN</sub>	Shutdown Supply Current	Iown Supply Current SHDN=0V, (Switch OFF)		20	35	μA
I <sub>IN SHDN</sub>	Shutdown Input Current	SHDN = 0 to $V_{IN}$	0	15	100	nA
		SHDN = $V_{IN}$ to 5V		12	40	μA
υ	Efficiency	I <sub>LOAD</sub> =100mA	0	82		%
V <sub>IH</sub>	Shutdown Input Threshold	V <sub>IN</sub> =1V to 6V	١	/ <sub>IN</sub> /2 + 0.25	5	V
I <sub>LIM</sub>	Current Limit	S		1.0		А
ILIM TEMPCO	Current Limit Temperature Coefficient	00-		-0.3		%/°C
t <sub>OFFMIN</sub>	Minimum Switch Off Time			1.2		μs
t <sub>ONMAX</sub>	Maximum Switch ON Time	V <sub>IN</sub> =2.5V		4.5		
	Å	V <sub>IN</sub> =1.8V		6.5		
		V <sub>IN</sub> =1V		15		
V <sub>CESAT NPN</sub>	Switch saturation Voltage	I <sub>SW</sub> =400mA		0.25		V
		I <sub>SW</sub> =600mA		0.33		
		I <sub>SW</sub> =1000mA		0.5		
V <sub>CESAT NPN</sub>	Rectifier Forward Drop	I <sub>SW</sub> =400mA		0.18		V
16		I <sub>SW</sub> =600mA		0.22		
~O``		I <sub>SW</sub> =1000mA		0.4		
V <sub>FB</sub>	Error Comparator Trip Point	ST779, over operating input voltage (Note 6)		1.23±2%		V
I <sub>FB</sub>	FB Pin Bias Current	ST779, V <sub>FB</sub> =1.3V		50		nA
	Switch Off Leakage Current			0.1		μA
I <sub>LX</sub>	Rectifier Off Leakage Current			0.1		μA

**Note 1:** Output in regulation,  $V_{OUT} = V_{OUT}$  (nominal) ± 4%. **Note 2:** At hight  $V_{IN}$  to  $V_{OUT}$  differentials, the maximum load current is limited by the maximum allowable power dissipation in the package. **Note 3:** Start-up guaranteed under these load conditions.

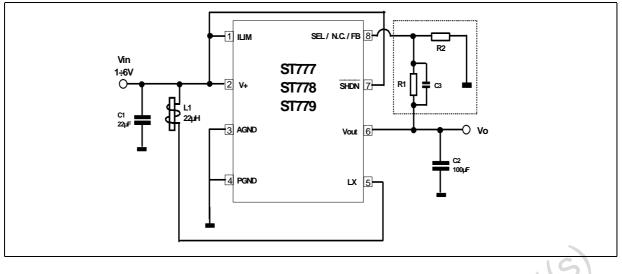
Note 4: Minimum value is production tested. Maximum value is guaranteed by design and is not production tested.

Note 5: In the ST779 supply current depends on the resistor divider used to set the output voltage.

Note 6:  $V_{OUT}$  is set to a target value of +5V by 0.1% external feedback resistors.  $V_{OUT}$  is measured to be 5V±2.5% to guarantee the error comparator trip point.



## **TYPICAL APPLICATION CIRCUIT**



## **APPLICATIONS INFORMATION**

R1 and R2 must be placed only in ST779 applications to set the output voltage according to the following equation:

 $V_{OUT} = (1.23) [(R1+R2)/R2]$ 

and to simplify the resistor selection:

 $R1 = R2 [(V_{OUT}/1.23)-1]$ 

It is possible to use a wide range of values for R2 ( $10K\Omega$  to  $50K\Omega$ ) with no significant loss of accuracy thanks to the very low FB input current. To have 1% error, the current through R2 must be at least 100 times FB's bias current.

When large values are used for the feedback resistors (R1> $50K\Omega$ ), stray output impedance at FB can incidentally add "lag" to the feedback response, destabilizing the regulator and creating a larger ripple at the output. Lead lengths and circuit board traces at the FB node should be kept short. Compensate the loop by adding a "lead" compensation capacitor (C3, 100pF to 1nF) in parallel with R1.

The typical value of the L1 inductor is  $22\mu$ H, enough for most applications. However, are also suitable values ranging from  $10\mu$ F to  $47\mu$ F with a saturation rating equal to or greater than the peak switch -current limit. Efficiency will be reduced if the inductor works near its saturation limit, while will be maximized using an inductor with a low DC resistance, preferably under  $0.2\Omega$ .

Connecting ILIM to  $V_{IN}$  the maximum LX current limit (1A) is set. If this maximum value is not required is possible to reduce it connecting a resistor between ILIM and  $V_{IN}$  (See Figure 16 to choose the right value). The current limit value is misured when the switch current through the inductor begins to flatten and does'nt coincide with the max short circuit current.

Even if the device is designed to tolerate a short circuit without any damage, it is strictly recommended to avoid a continuos and durable short circuit of the output to GND.

To achieve the best performances from switching power supply topology, particular care to layout drawing is needed, in order to minimize EMI and obtain low noise. Moreover, jitter free operation ensures the full device functionality. Wire lengths must be minimized, filter and by-pass capacitors must be low ESR type, placed as close as possible to the integrated circuit. Solder AGND and PGND pins directly to a ground plane.

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**TYPICAL CHARACTERISTICS** (unless otherwise specified  $T_i = 25^{\circ}C$ ,  $C_i = 22\mu$ F,  $C_O = 100\mu$ F)



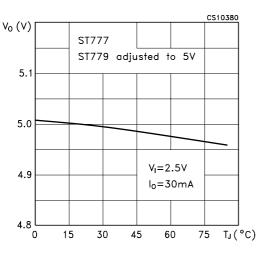
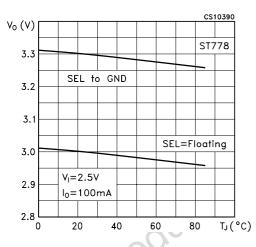
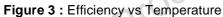
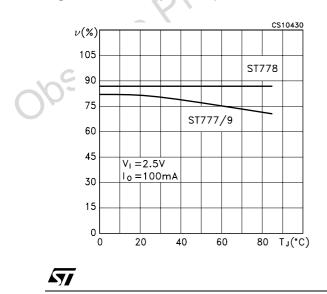
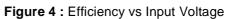


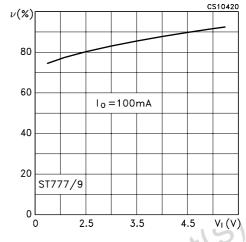
Figure 2 : Output Voltage vs Temperature

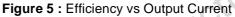












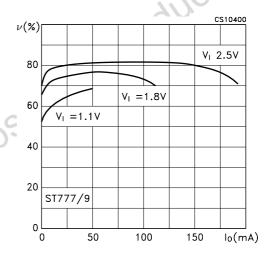


Figure 6 : Efficiency vs Low Output Current

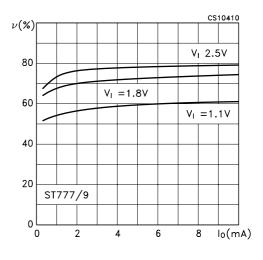
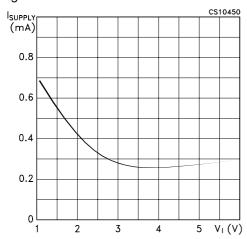
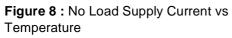
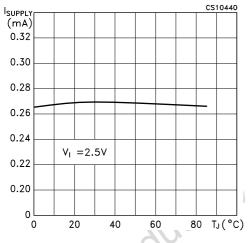
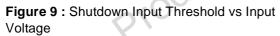


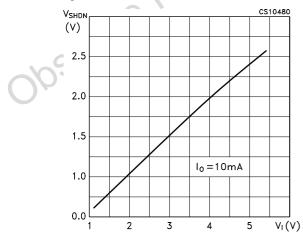
Figure 7 : No Load Supply Current vs Input Voltage

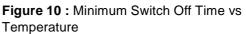


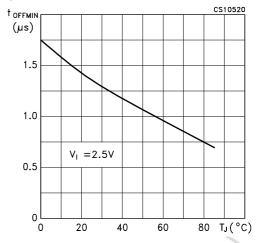


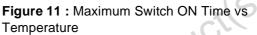












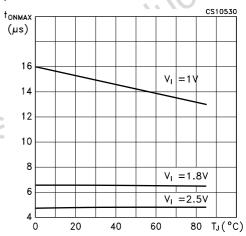


Figure 12 : FB Pin Bias Current vs Temperature

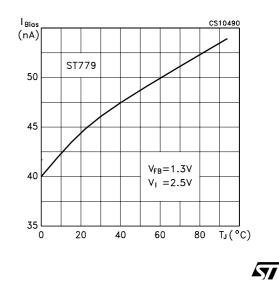
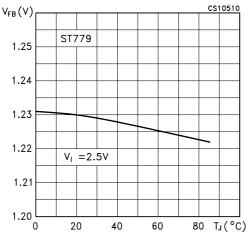
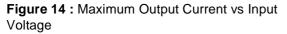


Figure 13 : Error Comparator Trip Point vs Temperature





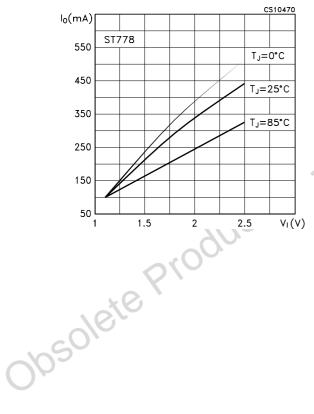
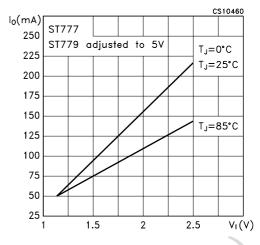
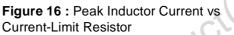
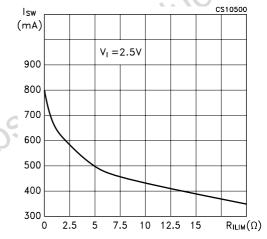
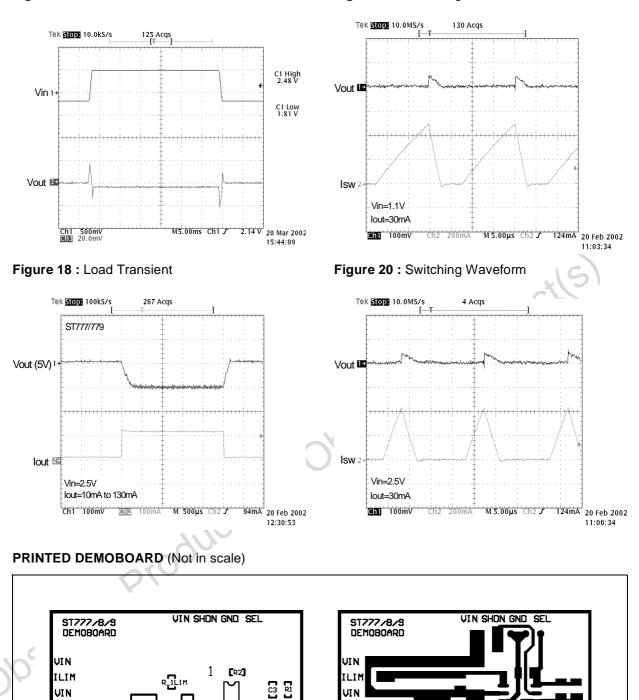


Figure 15 : Maximum Output Current vs Input Voltage









VIN

GND

VOUT

#### Figure 17 : Line Transient

Figure 19 : Switching Waveform

VIN

GND

VOUT

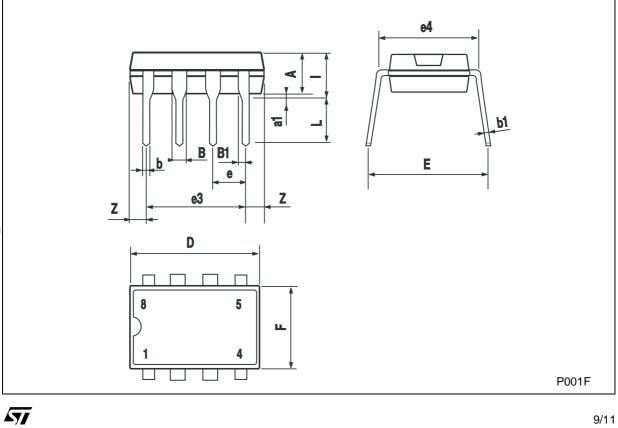
<u>C</u>1

[C2]

L1

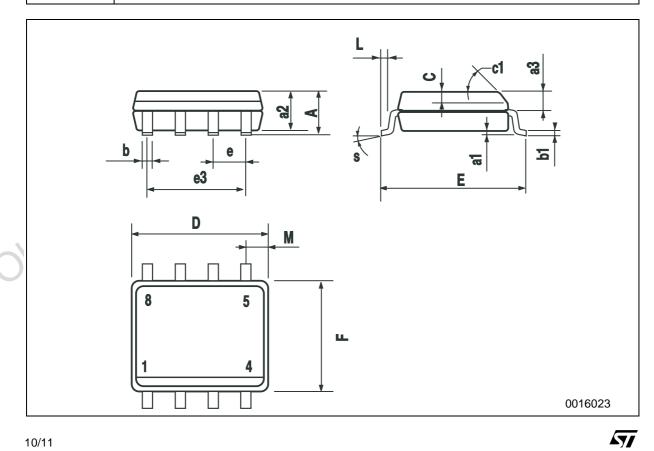
DIM.	mm.			inch		
	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.
А		3.3			0.130	
a1	0.7			0.028		
В	1.39		1.65	0.055		0.065
B1	0.91		1.04	0.036		0.041
b		0.5			0.020	
b1	0.38		0.5	0.015		0.020
D			9.8			0.386
E		8.8			0.346	
е		2.54			0.100	
e3		7.62			0.300	
e4		7.62			0.300	
F			7.1			0.280
I			4.8			0.189
L		3.3			0.130	

## **Plastic DIP-8 MECHANICAL DATA**



## ST777/778/779

DIM		mm.		inch			
DIM.	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.	
А			1.75			0.068	
a1	0.1		0.25	0.003		0.009	
a2			1.65			0.064	
a3	0.65		0.85	0.025		0.033	
b	0.35		0.48	0.013		0.018	
b1	0.19		0.25	0.007		0.010	
С	0.25		0.5	0.010		0.019	
c1		•	45°	(typ.)	•		
D	4.8		5.0	0.189		0.196	
E	5.8		6.2	0.228		0.244	
е		1.27			0.050		
e3		3.81			0.150		
F	3.8		4.0	0.149		0.157	
L	0.4		1.27	0.015		0.050	
М			0.6			0.023	



## SO-8 MECHANICAL DATA

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